Multiple-Access Channel with Delayed State Information Via Directed Information

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Basher/Shirazi/Permuter MAC with Delayed State Information Via Directed Information

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- Proof via directed information
- Capacity region for a finite state additive Gaussian MAC
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- Summary

 Instantaneous channel state information (CSI) at the transmitters is often an unrealistic assumption in wireless communications

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- The CSI can be transmitted to the transmitters through feedback
- But the CSI feedback is not instantaneous
- Motivated by this we studied the problem of Finite State Markov (FSM) Multiple-Access Channel (MAC) with delayed state information

Motivation-LTE



Channel Model



Figure: FSM-MAC with CSI at the decoder and delayed CSI at the encoders. We consider the above problem setting in the following cases: $d_1 > d_2$, $d_1 = d_2$, and $d_2 < d_1 = \infty$.

Strictly causal CSI [Steinberg/Lapidoth10] Point-to-point case [Viswanathan99]

- Finite number of states $\mathcal{S} < \infty$.
- Channel state is a stationary Markov process independent of the messages.
- The random variables S_i, S_{i-d} denote the channel state at time i, and i - d, respectively.
- The (S_i, S_{i-d}) joint distribution is stationary and is given by

$$P(S_i = s_l, S_{i-d} = s_j) = \pi(s_j)K^d(s_l, s_j).$$

• The channel transition probability at time *i* is given by $P(y_i|x_{1,i}, x_{2,i}, s_i)$.

For each encoder an encoding function,

$$X_{1,i} = \left\{ \begin{array}{cc} f_{1,i}(M_1), & 1 \le i \le d_1 \\ f_{1,i}(M_1, S^{i-d_1}), & d_1 + 1 \le i \le n \end{array} \right\}$$
(1)

$$X_{2,i} = \left\{ \begin{array}{cc} f_{2,i}(M_2), & 1 \le i \le d_2 \\ f_{2,i}(M_2, S^{i-d_2}), & d_2 + 1 \le i \le n \end{array} \right\}$$
(2)

Theorem

The capacity region of FSM-MAC with CSI at the decoder and asymmetrical delayed CSI at the encoders with delays d_1 and d_2 ($d_1 \ge d_2$), is given by:

$$\mathcal{R} = \bigcup_{P(u|\tilde{s}_1)P(x_1|\tilde{s}_1, u)P(x_2|\tilde{s}_1, \tilde{s}_2, u)} \begin{pmatrix} R_1 < I(X_1; Y|X_2, S, \tilde{S}_1, \tilde{S}_2, U) \\ R_2 < I(X_2; Y|X_1, S, \tilde{S}_1, \tilde{S}_2, U) \\ R_1 + R_2 < I(X_1, X_2; Y|S, \tilde{S}_1, \tilde{S}_2, U), \end{pmatrix}$$

where U is an auxiliary random variable with cardinality $|\mathcal{U}| \leq 3$.

The joint distribution $(S, \tilde{S}_1, \tilde{S}_2)$ is the same joint distribution as $(S_i, S_{i-d_1}, S_{i-d_2})$.

Theorem

The capacity region of FSM-MAC with CSI at the decoder and symmetrical delayed CSI at the encoders with delay d ($d = d_1 = d_2$), is given by:

$$\mathcal{R} = \bigcup_{P(u|\tilde{s})P(x_1|\tilde{s},u)P(x_2|\tilde{s},u)} \begin{pmatrix} R_1 < I(X_1; Y|X_2, S, \tilde{S}, U) \\ R_2 < I(X_2; Y|X_1, S, \tilde{S}, U) \\ R_1 + R_2 < I(X_1, X_2; Y|S, \tilde{S}, U), \end{pmatrix}$$

where U is an auxiliary random variable with cardinality $|\mathcal{U}| \leq 3$.

Theorem

The capacity region of FSM-MAC with CSI at the decoder and delayed CSI only to one encoder is given by :

$$\mathcal{R} = \bigcup_{P(q)P(x_1|q)P(x_2|\tilde{s},q)} \begin{pmatrix} R_1 < I(X_1; Y|X_2, S, \tilde{S}, Q) \\ R_2 < I(X_2; Y|X_1, S, \tilde{S}, Q) \\ R_1 + R_2 < I(X_1, X_2; Y|S, \tilde{S}, Q), \end{pmatrix}$$

where Q is an auxiliary random variable with cardinality $|Q| \leq 3$.

MAC with time-invariant feedback



[P/Weissman/Chen09]

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Definitions

$$I(X^{n};Y^{n}) = \sum_{i=1}^{n} I(X^{n};Y_{i}|Y^{i-1})$$

$$P(x^{n}|z^{n}) = \prod_{i=1}^{n} P(x_{i}|z^{n}, x^{i-1})$$

Definitions

Directed Information

[Massey90]

$$I(X^n \to Y^n) \triangleq \sum_{i=1}^n I(X^i; Y_i | Y^{i-1})$$
$$I(X^n; Y^n) = \sum_{i=1}^n I(X^n; Y_i | Y^{i-1})$$

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Causal conditioning

[Kramer98]

$$P(x^{n}||z^{n-d}) \triangleq \prod_{i=1}^{n} P(x_{i}|z^{i-d}, x^{i-1})$$
$$P(x^{n}|z^{n}) = \prod_{i=1}^{n} P(x_{i}|z^{n}, x^{i-1})$$

Finite State MAC

Let

$$\underline{\mathcal{R}}_{n} = \bigcup \begin{cases} R_{1} \leq \frac{1}{n} I(X_{1}^{n} \to Y^{n} || X_{2}^{n}, s_{0}), \\ R_{2} \leq \frac{1}{n} I(X_{2}^{n} \to Y^{n} || X_{1}^{n}, s_{0}), \\ R_{1} + R_{2} \leq \frac{1}{n} I((X_{1}, X_{2})^{n} \to Y^{n} |s_{0}), \end{cases}$$

the union is over input distribution $P(x_1^n || z_1^{n-1}) P(x_2^n || z_2^{n-1})$.

Theorem

For Finite state MAC with time invariant feedback and Markovian state, the capacity region is

$$\mathcal{R} = \lim_{n \to \infty} \mathcal{R}_n,$$

where the limit exists

P.& Weissman& Chen07

Proof via Directed information

- Adapt the feedback model:
 - State information at the decoder as a part of the channel's output
 - Choosing the deterministic function of the output:

$$z_{1,i}(y_i, s_i) = z_{2,i}(y_i, s_i) = s_i$$

• Multi-letter expression \rightarrow Single-letter expression.

$$\begin{array}{rcl} R_{1} &\leq & \frac{1}{n}I(X_{1}^{n} \to Y^{n}, S^{n} || X_{2}^{n}) \\ R_{1} &\leq & \frac{1}{n}I(X_{2}^{n} \to Y^{n}, S^{n} || X_{1}^{n}) \\ R_{1} + R_{2} &\leq & \frac{1}{n}I((X_{1}, X_{2})^{n} \to Y^{n}, S^{n}) \end{array}$$

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- Adapt the feedback model:
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• Multi-letter expression \rightarrow Single-letter expression.

$$R_{1} \leq \frac{1}{n}I(X_{1}^{n} \to Y^{n}, S^{n}||X_{2}^{n}) = I(X_{1}; Y|X_{2}, S, \tilde{S}_{1}, \tilde{S}_{2}, U),$$

$$R_{1} \leq \frac{1}{n}I(X_{2}^{n} \to Y^{n}, S^{n}||X_{1}^{n}) = I(X_{2}; Y|X_{1}, S, \tilde{S}_{1}, \tilde{S}_{2}, U),$$

$$R_{1} + R_{2} \leq \frac{1}{n}I((X_{1}, X_{2})^{n} \to Y^{n}, S^{n}) = I(X_{1}, X_{2}; Y|S, \tilde{S}_{1}, \tilde{S}_{2}, U).$$

FS additive Gaussian noise (AGN) MAC,

$$Y_i = X_{1,i} + X_{2,i} + N_{S_i}, \tag{3}$$

we apply the sum rate formula on the finite state Markov AGN MAC with transmitters power constrains \mathcal{P}_1 and \mathcal{P}_2 .

$$R_1 + R_2 < \max_{p(u|\tilde{s}_1)p(x_1|\tilde{s}_1, u)p(x_2|\tilde{s}_1, \tilde{s}_2, u)} I(X_1, X_2; Y|S, \tilde{S}_1, \tilde{S}_2, U),$$

subject to the power constraints,

$$\sum_{\tilde{s}_1} \pi(\tilde{s}_1) \sum_{u} P(u|\tilde{s}_1) E[X_1^2|\tilde{s}_1, u] \le \mathcal{P}_1,$$

$$\sum_{\tilde{s}_1} \pi(\tilde{s}_1) \sum_{\tilde{s}_2} P(\tilde{s}_2|\tilde{s}_1) \sum_{u} P(u|\tilde{s}_1) E[X_2^2|\tilde{s}_1, \tilde{s}_2, u] \le \mathcal{P}_2.$$

We bound the sum rate,

$$I(X_{1}, X_{2}; Y|S, \tilde{S}_{1}, \tilde{S}_{2}, U) \leq \frac{1}{2} \sum_{\tilde{s}_{1}} \pi(\tilde{s}_{1}) \sum_{\tilde{s}_{2}} P(\tilde{s}_{2}|\tilde{s}_{1}) \sum_{s} P(s|\tilde{s}_{2}) \\ \times \log\left(1 + \frac{\mathcal{P}_{1}(\tilde{s}_{1}) + \mathcal{P}_{2}(\tilde{s}_{1}, \tilde{s}_{2})}{\sigma_{s}^{2}}\right).$$
(4)

We can achieve (4) if we choose:

•
$$X_1(\tilde{s}_1, u) \sim \mathcal{N}(0, \mathcal{P}_1(\tilde{s}_1)).$$

•
$$X_2(\tilde{s}_1, \tilde{s}_1, u) \sim \mathcal{N}(0, \mathcal{P}_2(\tilde{s}_1, \tilde{s}_2))$$

•
$$X_1(\tilde{s}_1, u) \perp X_2(\tilde{s}_1, \tilde{s}_1, u) \perp N_s$$

We get the following optimization problem,

$$R_{1} + R_{2} = \max_{\mathcal{P}_{1}(\tilde{s}_{1}), \mathcal{P}_{2}(\tilde{s}_{1}, \tilde{s}_{2})} \frac{1}{2} \sum_{\tilde{s}_{1}} \pi(\tilde{s}_{1}) \sum_{\tilde{s}_{2}} K^{d_{1}-d_{2}}(\tilde{s}_{2}, \tilde{s}_{1}) \sum_{s} K^{d_{2}}(s, \tilde{s}_{2}) \\ \times \log\left(1 + \frac{\mathcal{P}_{1}(\tilde{s}_{1}) + \mathcal{P}_{2}(\tilde{s}_{1}, \tilde{s}_{2})}{\sigma_{s}^{2}}\right),$$
(5)

subject to the power constraints,

$$\sum_{\tilde{s}_{1}} \pi(\tilde{s}_{1}) \mathcal{P}_{1}(\tilde{s}_{1}) \leq \mathcal{P}_{1},$$

$$\sum_{\tilde{s}_{1}} \pi(\tilde{s}_{1}) \sum_{\tilde{s}_{2}} P(\tilde{s}_{2}|\tilde{s}_{1}) \mathcal{P}_{2}(\tilde{s}_{1},\tilde{s}_{2}) \leq \mathcal{P}_{2}.$$
(7)

Gilbert-Elliot Gaussian MAC

• At any given time *i* the channel is in one of two possible states *Good* or *Bad*.



Figure: Two-state AGN channel.

Gilbert-Elliot Gaussian MAC

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•
$$\sigma_B^2 > \sigma_G^2$$
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Gilbert-Elliot Gaussian MAC

• At any given time *i* the channel is in one of two possible states *Good* or *Bad*.

•
$$\sigma_B^2 > \sigma_G^2$$
.

• $\mathcal{P}_1 = 10, \mathcal{P}_2 = 10, \sigma_G^2 = 1, \sigma_B^2 = 100, g = 0.1, b = 0.1.$



Figure: Two-state AGN channel.

Two State AGN MAC- Sum Rate



Figure: Sum rate versus delay d (symmetrical delay $d_1 = d_2 = d$).

Two State AGN MAC- Power Control Policy



Figure: Power control policy versus delay d (symmetrical delay $d_1 = d_2 = d$).

Two State AGN MAC- Capacity Region



Figure: Capacity rate region for the two states AGN-MACsymmetrical case $d = d_1 = d_2$.

Two State AGN MAC- Capacity Region



Figure: Capacity rate region for the two states AGN-MAC-Transmitter 1 doesn't have the CSI $d_2 \le d_1 = \infty$.

Power Constrained FS Multiple-Access Fading Channel

FS multiple-access fading channel,

$$Y_i = h_1(s_i)X_{1,i} + h_2(s_i)X_{2,i} + N_{S_i},$$
(8)



Figure: The fading channel.

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Figure: The fading channel.

- The terms $X_{1,i}, X_{2,i}$ are the transmitted waveform.
- The terms $h_1(s_i), h_2(s_i)$ are the fading process of the users, and are deterministic functions of s_i .

We derive the following optimization problem,

$$R_{1} + R_{2} = \max_{\mathcal{P}_{1}(\tilde{s}_{1}), \mathcal{P}_{2}(\tilde{s}_{1}, \tilde{s}_{2})} \frac{1}{2} \sum_{\tilde{s}_{1}} \pi(\tilde{s}_{1}) \sum_{\tilde{s}_{2}} K^{d_{1}-d_{2}}(\tilde{s}_{2}, \tilde{s}_{1}) \\ \times \sum_{s} K^{d_{2}}(s, \tilde{s}_{2}) \log \left(1 + \frac{h_{1}(s)^{2} \mathcal{P}_{1}(\tilde{s}_{1}) + h_{2}(s)^{2} \mathcal{P}_{2}(\tilde{s}_{1}, \tilde{s}_{2})}{\sigma_{s}^{2}}\right),$$

subject to the power constraints,

$$\sum_{\tilde{s}_1} \pi(\tilde{s}_1) \mathcal{P}_1(\tilde{s}_1) \leq \mathcal{P}_1,$$

$$\sum_{\tilde{s}_1} \pi(\tilde{s}_1) \sum_{\tilde{s}_2} P(\tilde{s}_2|\tilde{s}_1) \mathcal{P}_2(\tilde{s}_1, \tilde{s}_2) \leq \mathcal{P}_2.$$

The state process,



The fading process,

$$h_1 = \left\{ \begin{array}{cc} 1 & s = 1 \\ 0.5 & s = 2 \end{array} \right\}, \quad h_2 = \left\{ \begin{array}{cc} 0.5 & s = 1 \\ 1 & s = 2 \end{array} \right\}.$$

- The noise, $N \rightarrow N(0, 1)$
 - $N \sim \mathcal{N}(0, 1).$
- The power constraints,

$$\mathcal{P}_1 = 2, \, \mathcal{P}_2 = 2.$$

Power Constrained FS Multiple-Access Fading Channel



Figure: Capacity rate region for the two states fading channelsymmetrical case $d = d_1 = d_2$.

Power Constrained FS Multiple-Access Fading Channel



Figure: Capacity rate region for the two states fading channel-Transmitter 1 doesn't have the CSI $d_2 \le d_1 = \infty$.

Summary

 A single-letter characterization is provided via directed information for the capacity region of FSM-MAC, when the transmitters have access to delayed CSI, and CSI is available at the receiver.

$$\bigcup_{P(u|\tilde{s}_1)P(x_1|\tilde{s}_1,u)P(x_2|\tilde{s}_1,\tilde{s}_2,u)} \begin{pmatrix} R_1 < I(X_1;Y|X_2,S,\tilde{S}_1,\tilde{S}_2,U) \\ R_2 < I(X_2;Y|X_1,S,\tilde{S}_1,\tilde{S}_2,U) \\ R_1 + R_2 < I(X_1,X_2;Y|S,\tilde{S}_1,\tilde{S}_2,U), \end{pmatrix}$$

- the whole statistics $(S, \tilde{S}_1, \tilde{S}_2)$ is important
- Computable result on FS AGN MAC, and on FS multiple-access fading channel.

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 A single-letter characterization is provided via directed information for the capacity region of FSM-MAC, when the transmitters have access to delayed CSI, and CSI is available at the receiver.

$$\bigcup_{P(u|\tilde{s}_1)P(x_1|\tilde{s}_1,u)P(x_2|\tilde{s}_1,\tilde{s}_2,u)} \begin{pmatrix} R_1 < I(X_1;Y|X_2,S,\tilde{S}_1,\tilde{S}_2,U) \\ R_2 < I(X_2;Y|X_1,S,\tilde{S}_1,\tilde{S}_2,U) \\ R_1 + R_2 < I(X_1,X_2;Y|S,\tilde{S}_1,\tilde{S}_2,U), \end{pmatrix}$$

- the whole statistics $(S, \tilde{S}_1, \tilde{S}_2)$ is important
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Thank you to Benjamin Zaidel and thank you for attending the talk!!

Encoder 1:

- Construct k codebooks C_{š1} for all Š₁ ∈ S, when in each codebook C_{š1} there are 2^{n1(š1)R1(š1)} codewords.
- Every codeword $C_{\tilde{s}_1}(i)$ when $i \in \{1, 2, ..., 2^{n_1(\tilde{s}_1)R_1(\tilde{s}_1)}\}$ has a length of $n_1(\tilde{s}_1)$ symbols.
- Each codeword from the $C_{\tilde{s}_1}$ codebook is built $X_{\tilde{s}_1} \sim \text{i.i.d.}$ $P(x_{\tilde{s}_1}|\tilde{S}_1 = \tilde{s}_1)$.
- Every time that the delayed CSI is S₁ = s₁, encoder 1 sends the next symbol from C_{s1} codebook.
- Therefore we can send total of $2^{nR_1} = 2^{\sum_{\tilde{s}_1 \in S} n_1(\tilde{s}_1)R_1(\tilde{s}_1)}$ messages.

Two State AGN MAC- Capacity Region

We present the capacity rate region by solving numerically the following optimization problem for different values of α ,

$$\max_{R_1,R_2} \alpha R_1 + R_2,$$
 (9)

subject to the constraints,

$$R_{1} \leq \frac{1}{2} \sum_{\tilde{s}_{1}} \pi(\tilde{s}_{1}) \sum_{\tilde{s}_{2}} K^{d_{1}-d_{2}}(\tilde{s}_{2}, \tilde{s}_{1}) \\ \times \sum_{s} K^{d_{2}}(s, \tilde{s}_{2}) \log\left(1 + \frac{\mathcal{P}_{1}(\tilde{s}_{1})}{\sigma_{s}^{2}}\right), \quad (10)$$

$$R_{2} \leq \frac{1}{2} \sum_{\tilde{s}_{1}} \pi(\tilde{s}_{1}) \sum_{\tilde{s}_{2}} K^{d_{1}-d_{2}}(\tilde{s}_{2}, \tilde{s}_{1}) \\ \times \sum_{s} K^{d_{2}}(s, \tilde{s}_{2}) \log\left(1 + \frac{\mathcal{P}_{2}(\tilde{s}_{1}, \tilde{s}_{2})}{\sigma_{s}^{2}}\right), \quad (11)$$

$$R_{1} + R_{2} \leq \frac{1}{2} \sum_{\tilde{s}_{1}} \pi(\tilde{s}_{1}) \sum_{\tilde{s}_{2}} K^{d_{1}-d_{2}}(\tilde{s}_{2}, \tilde{s}_{1}) \sum_{s} K^{d_{2}}(s, \tilde{s}_{2}) \\ \times \log\left(1 + \frac{\mathcal{P}_{1}(\tilde{s}_{1}) + \mathcal{P}_{2}(\tilde{s}_{1}, \tilde{s}_{2})}{\sigma_{s}^{2}}\right). \quad (12)$$

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$$R_{1} + R_{2} = \max_{\mathcal{P}_{1}(\tilde{s}), \mathcal{P}_{2}(\tilde{s})} \frac{1}{2} \sum_{\tilde{s}} \pi(\tilde{s}) \sum_{s} K^{d}(s, \tilde{s}) \log\left(1 + \frac{\mathcal{P}_{1}(\tilde{s}) + \mathcal{P}_{2}(\tilde{s})}{\sigma_{s}^{2}}\right),$$
(13)

subject to the power constraints,

$$\sum_{\tilde{s}} \pi(\tilde{s}) \mathcal{P}_1(\tilde{s}) \le \mathcal{P}_1, \tag{14}$$

$$\sum_{\tilde{s}} \pi(\tilde{s}) \mathcal{P}_2(\tilde{s}) \le \mathcal{P}_2, \tag{15}$$

$$\mathcal{P}_1(\tilde{s}) \ge 0 \quad \forall \tilde{s},\tag{16}$$

$$\mathcal{P}_2(\tilde{s}) \ge 0 \quad \forall \tilde{s}. \tag{17}$$

The solution can be obtained by the Lagrange multiplier method. Since \log is concave function, and that $\pi(\tilde{s}), K^d(s, \tilde{s}) \ge 0$. We get that objective function is concave in both variables $\mathcal{P}_1(\tilde{s})$, and $\mathcal{P}_2(\tilde{s})$. Also the constraints function are affine. So we can use the Kuhn-Tucker conditions as a sufficient conditions to solve the optimization problem.